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Abstract—A compact set of equations based on the multiple subbands quasi-ballistic transport theory is developed, and is used to investigate the channel electron effective mobility in recently reported In_{0.53}Ga_{0.47}As/Al₂O₃ tri-gate n-FinFET. The extracted electron effective mobility μ_n is around 370 cm²/V·s at low $V_g - V_{th}$ bias at room temperature and decreases with increasing V_g , and increases with increasing temperature (240–332K). It is very different from the case of Si n-MOSFETs, where the electron mobility decreases with increasing temperature. The low channel effective mobility and the ab-normal temperature dependence of μ_n are ascribed to the high acceptor interface trap and border trap energy densities in the conduction band energy of InGaAs. The ballistic channel resistance R_{Ball} at low V_{ds} is calculated and compared with the measured channel resistance R_{CH} . The low transmission coefficient $T = R_{Ball}/R_{CH} \approx 0.06$ to 0.05 indicates that there is a large room to improve the InGaAs/Al₂O₃ n-FinFET performance.

Index Terms—Al₂O₃, FinFETs, InGaAs, mobility, nano scale transistor, quasi-ballistic transport.

I. INTRODUCTION

INGAAS has high bulk electron mobility [1] and the potential to replace the conventional Si for the channel material in n-MOSFETs for low power low voltage logic applications [2]–[4]. However, the channel mobility in MOSFET is very different from the bulk mobility [5]. Particularly, it is size and structural dependent when the device is scaled down to the nanosize, due to the more severe effect related to the surface or interface scattering mechanisms [6], [7]. Therefore, it is highly desirable (however is lack of existing work) on reliable channel mobility extraction directly from nanoscale InGaAs n-MOSFETs. In this study, based on the quasi-ballistic transport theory of nano MOSFETs by Lundstrom's group [8], [9], we have developed a compact set of equations for physical characterization of quasi-ballistic transport and channel electron effective mobility μ_n in

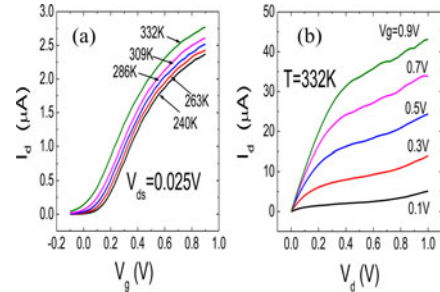


Fig. 1. (a) I_d - V_g curves of In_{0.57}Ga_{0.47}As n-FinFET with channel length $L = 100$ nm, Fin width $W_{Fin} = 40$ nm, Fin Height $H_{Fin} = 40$ nm, at different temperature. (b) I_d - V_d curves of the same device. The curves are measured by Agilent 4156 and after source-drain resistance correction [10].

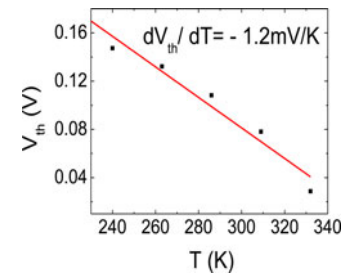


Fig. 2. V_{th} versus temperature of In_{0.57}Ga_{0.47}As n-FinFET with $L = 100$ nm. $dV_{th}/dT \approx -1.2$ mV/T.

the In_{0.53}Ga_{0.47}As n-FinFET with fin width $W_{Fin} = 40$ nm, fin height $H_{Fin} = 40$ nm, and channel length $L = 100$ nm, and with 5 nm ALD Al₂O₃ gate dielectric thickness. The detailed device structure and fabrication process were described in [4].

II. EXPERIMENTAL DATA AND THE MOBILITY EXTRACTION ALGORITHM

Figs. 1 and 2 show the experimental I_d - V_g and I_d - V_d curves and the temperature dependence of the threshold voltage V_{th} of the In_{0.53}Ga_{0.47}As n-FinFET, measured by Agilent 4156 and after source-drain resistance correction [10] in the temperature range of 240–332 K.

For one subband electrons, the I_d - V_g equation in the low V_{ds} region in [8] can be expressed by

$$I_d \cong \frac{1}{L} \left[\frac{1}{\mu_n} + \frac{1}{\mu_{Ballistic}} \right]^{-1} Q_n V_{ds} \quad (1)$$

Different from the conventional 2-D area charge density description as in [1], [8], in (1), we introduce the 1-D line density of mobile charge Q_n along the channel direction at the top of the source-channel barrier. μ_n and μ_{Ball} are the diffusive mobility and ballistic mobility, respectively. In the Boltzmann distribution case, from the well-known relationship between

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the electron mean free path λ_n and the mobility μ_n and the diffusion coefficient D_n derived by Shockley [11], the diffusive

mobility:

$$\mu_n = \frac{q}{k_B T} D_n = \frac{q}{k_B T} \left[\frac{v_T \lambda_n}{2} \right]. \quad (2a)$$

Here q is the absolute value of the electron charge, v_T is the electron thermal velocity

$$v_T = \sqrt{\frac{2k_B T}{\pi m_n^*}}$$

where m_n^* is the electron effective mass of In_{0.53}Ga_{0.47}As (= 0.043 m_0) [12]. Extend to the Fermi–Dirac distribution for one subband case, the following equation can be obtained [8]:

$$\mu_n = \frac{qv_T \lambda_n}{2k_B T} \left[\frac{F_{-1/2}(\eta_F)}{F_0(\eta_F)} \right]. \quad (2)$$

Here, $F_j(\eta_F)$ is the j th-order F-D integral [13], $\eta_F = (E_F - E_s)/k_B T$, E_s is the bottom energy of the subband.

The ballistic mobility μ_{Ball} is first introduced by Shur in the Boltzmann distribution case [14], and later extended to the Fermi–Dirac distribution case by Lundstrom *et al* [8]

$$\mu_{\text{Ballistic}} = \frac{qv_T L}{2k_B T} \left[\frac{F_{-1/2}(\eta_F)}{F_0(\eta_F)} \right]. \quad (3)$$

The following two issues should be carefully considered for reliable mobility extraction.

- 1) In In_{0.53}Ga_{0.47}As n-FinFET with 40-nm fin width, the quantum confinement induced energy lifts for the electron subbands $E_c - E_s^i$ are very small (quasi-classic). E_s^i is the bottom energy of the i th subband. For a very rough approximation of an infinitely high barrier square quantum well, $E_c - E_s^1$ is only $(\pi\hbar)^2/2m_n^* W_{\text{Fin}}^2 = 5.5$ meV for the first subband. When increasing V_g , the number of occupied subbands increases and (1) to (3) should be extended to include multiple subbands.
- 2) The charge density Q_n cannot be estimated by the conventional experimental method by measuring the gate capacitance versus V_g [5]. The gate capacitance cannot be accurately measured because of the high density of interface traps induced dispersion [15] and very small gate area of the FinFET, and can only be estimated by careful simulation. The 1-D line density of gate capacitance $C_G = dQ_n/dV_g$ corresponding to the 1-D charge density Q_n is introduced. If the low doping acceptor charge in the fin body under volume inversion is neglected, C_G consists of two components in series. The first component is the oxide capacitance 1-D line density

$$C_{\text{OX}} = \frac{\kappa\epsilon_0}{t_{\text{ox}}} (W_{\text{Fin}} + 2H_{\text{Fin}}) = (14.2 \text{ fF}/\mu\text{m}^2) (W_{\text{Fin}} + 2H_{\text{Fin}}) \quad (4a)$$

where $\kappa = 8$ is the dielectric constant of Al₂O₃ [16], ϵ_0 is the permittivity of vacuum, and $t_{\text{ox}} = 5$ nm is the physical thickness of the dielectric. For Fin body with $W_{\text{Fin}} = H_{\text{Fin}} = 40$ nm \gg dielectric thickness 5 nm, (4a) is a good approximation for C_{OX} .

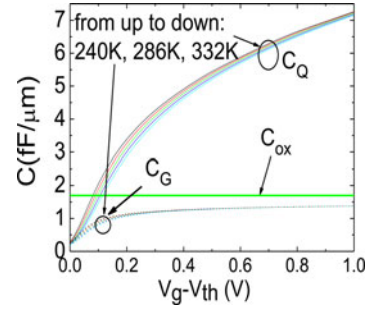


Fig. 3. Simulation results of 1D C_Q and C_G versus $V_g - V_{\text{th}}$ at different temperature by using (11)–(13). The gate oxide 1-D capacitance C_{OX} (5 nm thick Al₂O₃ with dielectric constant = 8, Fin width $W_{\text{Fin}} = 40$ nm, Fin Height $H_{\text{Fin}} = 40$ nm) is also shown as reference.

The second component is the quantum capacitance (inversion capacitance) 1-D line density $C_Q = dQ_n/dV_Q$ here dV_Q is the Fin inversion potential change under volume inversion induced by the change of gate voltage dV_g . According to [17] and under volume inversion, we can estimate the one subband quantum capacitance 1-D line density to be

$$C_Q (\text{one subband}) = (q^2 m_n^* / \pi \hbar^2) H_{\text{Fin}} = (28.7 \text{ fF}/\mu\text{m}^2) H_{\text{Fin}} \quad (4b)$$

which is comparable with the value of C_{OX} . The multiple subband quantum capacitance C_Q is a function of V_g (see (12) and Fig. 3), also comparable with C_{OX} .

Keeping points (I) and (II) in mind, we have extended (1)–(3) to a generic and compact set of (5)–(13) for multiple subbands quasi-ballistic transport in nanoscale FinFETs (or nanowire MOSFETs) in the linear V_{ds} region.

According to the similar method in [8], the 1-D Q_n can be expressed by

$$Q_n = \sum_i Q_i \approx q N_{2D} H_{\text{Fin}} \sum_i F_0(\eta_F^i) \quad (5)$$

$$N_{2D} = \frac{m_n^* k_B T}{\pi \hbar^2}. \quad (5a)$$

Extending I_d in (1) to the multiple subbands case in the low V_{ds} region, and combining (2), (3), and (5)

$$\begin{aligned} I_d &\cong \frac{1}{L} \sum_i \left[\frac{1}{\mu_n^i} + \frac{1}{\mu_{\text{Ballistic}}^i} \right]^{-1} Q_i V_{ds} \\ &\approx \left[\frac{H_{\text{Fin}}}{L} \right] \left[\frac{qv_T}{2kT} \right] \left[\frac{1}{\lambda_n} + \frac{1}{L} \right]^{-1} V_{ds} q N_{2D} \sum_{i=1}^n F_{-1/2}(\eta_F^i). \end{aligned} \quad (6)$$

Introducing the 1-D channel resistance R_{CH} and using (5)

$$\begin{aligned} R_{\text{CH}} &= \left[\frac{V_{ds}}{I_d} \right]_{V_{ds} \rightarrow 0} = L \left[\frac{1}{Q_n \mu_{\text{Ball}}} + \frac{1}{Q_n \mu_n} \right] \\ &= R_{\text{Ball}} + R_{\text{diffusive}}. \end{aligned} \quad (7)$$

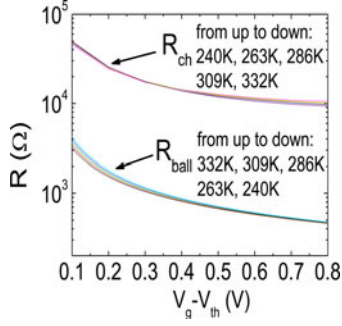


Fig. 4. Calculated R_{Ball} and measured R_{CH} for $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$ n-FinFET with channel length $L = 100$ nm, Fin width $W_{\text{Fin}} = 40$ nm, Fin Height $H_{\text{Fin}} = 40$ nm. The transmission coefficient $T = R_{\text{Ball}}/R_{\text{CH}}$ is around 0.06–0.05 for these curves.

Here the ballistic and channel mobilities are

$$\mu_{\text{Ball}} = \frac{qv_T L}{2k_B T} \left[\frac{\sum_i F_{-1/2}(\eta_F^i)}{\sum_i F_0(\eta_F^i)} \right] \quad (8)$$

$$\mu_n = \frac{qv_T \lambda_n}{2k_B T} \left[\frac{\sum_i F_{-1/2}(\eta_F^i)}{\sum_i F_0(\eta_F^i)} \right]. \quad (9)$$

Here $\eta_F^i = (E_F - E_s^i)/k_B T$. (7) indicates that the channel resistance R_{CH} consists of two components in series. In the ballistic limit when $L \ll \lambda_n$, there is only the first term R_{Ball} with ballistic electron mobility μ_{Ball} . R_{Ball} is independent of L , originated from the quantum contact resistance [18]. In the diffusive limit when $L \gg \lambda_n$, there is only the second term $R_{\text{diffusive}}$ with conventional channel electron mobility μ_n , and (7) is reduced to the long channel I_d - V_g equation [1]. Q_n and C_Q can be obtained by (10) to (13):

$$Q_n = \int_{V_g} C_G(V) dV = \frac{qm_n^* k T H_{\text{Fin}}}{\pi \hbar^2} \sum_i F_0(\eta_F^i) \quad (10)$$

$$C_G(V) = \frac{C_{\text{ox}} C_Q}{C_{\text{ox}} + C_Q} \quad (11)$$

$$C_Q = H_{\text{Fin}} \sum_i \frac{q^2 m_n^*}{\pi \hbar^2} F_{-1}(\eta_F^i). \quad (12)$$

The relationship between V_g and η_F^1 can be obtained by the following consideration. Since a change in the gate voltage dV_g induces a change in the Fin inversion potential dV_Q by $dV_g/dV_Q = C_Q/C_G$, we can derive

$$V_g(\eta_F^1) - V_{\text{th}} = \int_{\eta_F^1}^{\eta_F^1} \frac{C_Q k T}{q C_G} d\eta_F \quad (13)$$

where η_F^1 in (13) is the η_F^1 value corresponding to $V_g = V_{\text{th}}$, and can be estimated by (10) with mobile charge area density $= 10^{11} \text{ cm}^{-2}$ [19]. The calculated C_G versus V_g is shown in Fig. 3.

The calculated results of R_{Ball} in (7) are shown in Fig. 4. It is the typical ballistic limit of the channel resistance of $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$ n-FinFET. The measured channel resistance $R_{\text{CH}} = [V_{ds}/I_d]_{V_{ds} \rightarrow 0}$ obtained from the experimental data of I_d - V_g curves in Fig. 1 is also shown in Fig. 4 for comparison.

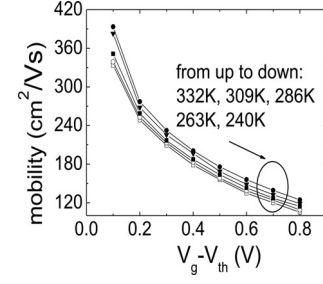


Fig. 5. Electron channel effective mobility μ_n of $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$ n-FinFET with $L = 100$ nm at different temperature 240–332 K versus $V_g - V_{\text{th}}$.

The transmission coefficient T of the flow of electrons [18] in low V_{ds} can be expressed by

$$T = \frac{\lambda_n}{\lambda_n + L} = \frac{R_{\text{Ball}}}{R_{\text{CH}}} \quad (14)$$

where T is related to the backscattering (or reflection) coefficient r by $T = (1-r)$ [18], [8], [9]. From the curves in Fig. 4, T is estimated to be around 0.06–0.05 only. It indicates that there is still a large room to be improved for the present fabrication technology, as will be further discussed later.

III. DISCUSSION

From the data in Fig. 4 and Q_n calculated by using (10)–(13), the channel electron mobility μ_n can be obtained from (7). The results are plotted in Fig. 5. There are two points worth to be mentioned for the results in Fig. 5. 1) The overall channel mobility μ_n in the $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$ n-FinFET is quite low. 2) The μ_n increases with increasing temperature, in contrast with the case in Si planar n-MOSFETs where μ_n decreases with increasing temperature at moderate surface field due to the phonon scattering [5].

These ab-normal mobility characteristics are ascribed to the high density of acceptor like interface traps [15], [20] and border traps [21], [22] in the conduction band of $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$ in the $\text{InGaAs}/\text{Al}_2\text{O}_3$ n-MOSFETs. The high densities of acceptor interface traps and border traps may response to the charge density estimated by (10) and (13). As a result, the real mobile charge density and the measured I_d are reduced. In the discussion of high-k/metal gate Si n-MOSFETs, Zhu *et al.* [23] have developed a method of correction to estimate the real mobile charge density out of high density of interface trap charge. The real mobility extracted by their method after correction is much higher than the effective mobility extracted by the conventional method [5], [24]. However, this higher real mobility actually is not beneficial to higher I_d . In our work, we would rather follow the conventional mobility studies [5], [24], [25] and ascribe the degraded effective mobility extracted by (7) to (13) to the existence of high density interface traps and border traps [26]. The charged interface traps and border traps not only reduce the mobile charge density at fixed V_g , but also induce Coulomb scattering associated with a Coulomb scattering mobility component μ_{Coulomb} , which increase with increasing temperature [1]. It is consistent with our experimental result. The results shown in Fig. 4 indicate that there is a large room to improve the present

InGaAs n-FinFET fabrication technology. The first way to improve the transmission coefficient T is obviously to reduce the channel length L . The more difficult task to improve T is to improve the InGaAs/dielectric interface to reduce the acceptor like interface trap and border trap energy densities in the InGaAs conduction band [15], [20], [21], [22]. On the other hand, the scattering mechanism and the ab-normal temperature dependence of effective mobility μ_n need more deep investigation and understanding to further improve the electric performance of the nano-InGaAs n-FinFETs.

IV. CONCLUSION

The quasi-ballistic transport and the effective electron mobility μ_n of In_{0.53}Ga_{0.47}As n-FinFET are investigated, based on the multiple subband quasi-ballistic theory and the experimental measurements on the FinFET devices with 100-nm channel length reported in [4]. The extracted μ_n is quite low, around 370 cm²/V·s at room temperature at low $V_g - V_{th}$ bias and decreases with increasing V_g , and increases with increasing temperature (240–332 K), in contrast with the case of planar Si n-MOSFET where μ_n decreases with increasing temperature. The ballistic channel resistance R_{Ball} at low V_{ds} is calculated and compared with the measured channel resistance R_{CH} . The ab-normal mobility characteristic is ascribed to high densities of acceptor interface traps and border traps in the energy range of InGaAs conduction band. The low transmission coefficient $T = R_{Ball}/R_{CH} \approx 0.06$ to 0.05 indicates that there is large performance room to be improved when the InGaAs/dielectric interface traps and border traps can be significantly reduced.

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